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# Timing jitter in passively mode-locked semiconductor lasers

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#### Abstract

We study the effect of noise on the dynamics of passively mode-locked semiconductor lasers both experimentally and theoretically. A method combining analytical and numerical approaches for estimation of pulse timing jitter is proposed. We investigate how the presence of dynamical features such as wavelength bistability affects timing jitter.

## 1 Introduction

Semiconductor mode-locked lasers received much attention in the last decade due to their multiple potential applications including high speed optical telecommunications and clocking [2, 3]. Pulses generated by these lasers are affected by noise due to spontaneous emission and other factors such as cavity optical length fluctuations. In particular, their temporal positions in a pulse train deviate from those of the perfectly periodic output. This phenomenon called timing jitter limits the performance of mode-locked devices [9]. Passive mode-locking is an attractive technique for periodic short pulse generation due to simplicity of implementation and handling as compared to other techniques such as hybrid or active mode-locking. However, in the absence of external reference clock passively mode-locked lasers exhibit relatively large pulse timing jitter [11]. Bistable semiconductor lasers can be used as a basic element for optical switches [15, 17] where the timing jitter can play a significant role as well.

Analytical approach to the study of the influence of noise on mode-locked pulses propagating in the laser cavity was developed by H. Haus and A. Mecozzi [7]. Later this technique was extended to include the effects of carrier density in semiconductor lasers [8]. However, many simplifications involved in the analysis of Refs. [7,8] limit its applicability to modelling dynamics of semiconductor lasers under the influence of noise. In the last two decades extensive numerical simulations of travelling wave [12, 22] and delay differential [13] models were performed to study timing jitter for different laser device configurations. In particular, a monotonous decrease of pulse timing jitter with the increase of injection current in a passively mode-locked semiconductor laser was demonstrated both numerically and experimentally [22]. In this paper using a delay differential equation (DDE) model of a two-section passively mode-locked semiconductor laser [18-20] we study the effect of noise on the characteristics of fundamental mode-locked regime and develop a semianalytical method for estimation of pulse timing jitter, which helps us to avoid high computational cost of a purely numerical approach. With the help of DDE-BIFTOOL [4] we perform numerical bifurcation analysis of the model and demonstrate the existence of bistability between two fundamental mode-locked regimes with different pulse repetition frequencies. By varying the injection current applied to the gain section we demonstrate both experimentally and theoretically three types of the injection current/timing jitter dependence: monotonous decrease of the pulse timing jitter with the increasing injection current, peaks of timing jitter at certain intermediate values of injection current corresponding to bifurcation points, and abrupt drop of the timing jitter level after a transition between two stable branches of fundamental mode-locked regime occurs. We find both theoretically and experimentally, in some laser samples, timing jitter can be reduced significantly with the additional increase of the injection current, which also shifts the operation frequency.

### 2 Method

#### 2.1 Delay differential model

We consider a delay differential equation (DDE) model of a two-section passively mode-locked semiconductor laser introduced in [18–20]:

$$\partial_t A + \gamma A = \gamma \sqrt{\kappa} \exp\left\{\frac{1 - i\alpha_g}{2}G\left(t - \tau\right) - \frac{1 - i\alpha_q}{2}Q\left(t - \tau\right)\right\} A\left(t - \tau\right) + \xi \eta(t), \tag{1}$$

$$\partial_t G = g_0 - \gamma_g G - \left(e^G - 1\right) |A|^2, \tag{2}$$

$$\partial_t Q = q_0 - \gamma_q Q - s \left( 1 - e^{-Q} \right) e^G |A|^2 \,, \tag{3}$$

where A(t) is the electric field envelope, G(t) and Q(t) are saturable gain and loss introduced by the gain and absorber sections correspondingly,  $\tau$  is the cold cavity round-trip time,  $g_0$  and  $q_0$  are unsaturated gain (pump) and absorption parameters,  $\alpha_{g,q}$  are linewidth enhancement factors in the gain and absorber sections,  $\gamma_{g,q}$  are carrier relaxation rates,  $\gamma$  is the spectral filtering bandwidth, s is a saturation parameter,  $\kappa$  is the linear attenuation factor per cavity round trip, and  $\eta(t)$  is  $\delta$ -correlated Langevin noise of amplitude  $\xi$  [13]:

$$\eta(t) = \eta_1(t) + i\eta_2(t), \ \langle \eta_i(t) \rangle = 0, \ \langle \eta_i(t)\eta_j(t') \rangle = \delta_{i,j}\delta(t-t').$$

We denote by  $\psi(t)=(\operatorname{Re} A,\operatorname{Im} A,G,Q)^T$  a real-valued solution of (1).

#### 2.2 Estimation of timing jitter

Due to time shift invariance of the autonomous system of equations (1)-(3) the timing phase of the solution  $\psi(t)$  exhibits a random walk under the influence of the white noise [12, 13]. This leads to the appearance of the pulse timing jitter. Hence, in order to estimate timing jitter one needs to separate the phase noise from the amplitude noise. In numerical simulations pulse timing jitter can be estimated directly by calculating the variance of the temporal pulse positions  $\sigma_{var}$  selected in a specific manner from the time series of the electric field A(t) [8,12]. Experimentally, the timing jitter is usually estimated using the root mean square (RMS),  $\sigma_{RMS}$ , of the power spectrum of the pulse train [12, 13, 22]. Estimation of the variance of temporal pulse positions  $\sigma_{var}$  can be done experimentally as well using a second-harmonic non-collinear optical cross-correlation technique, where the relation  $\sigma_{RMS} \approx 2.3548\sqrt{2}\sigma_{var}$  holds [8].

In this work we develop a semianalytical method for estimation of pulse timing jitter  $\sigma_{var}$  in DDE model (1)-(3) using a perturbation technique similar to the method proposed by H. Haus and A. Mecozzi [7]. We consider a locally stable fundamental mode-locked solution  $\psi_0(t) = (\operatorname{Re} A_0, \operatorname{Im} A_0, G_0, Q_0)^T$  of system (1)-(3) with  $\xi = 0$  (see Fig. 1(a)). This solution satisfies the conditions  $|A_0(t)|^2 = |A_0(t - T_0)|^2$ ,  $G_0(t) = G_0(t - T_0)$ , and  $Q_0(t) = Q_0(t - T_0)$ , where the mode-locked pulse period  $T_0$  is close to the cavity round trip time T. Now let us suppose that  $\xi \ll 1$ . The local dynamics of trajectories of equations (1)-(3) near the periodic solution can be studied using linearisation of the system. Substituting  $\psi(t) = \psi_0(t) + \delta\psi(t)$  into (1)-(3) and assuming that  $|\delta\psi| \ll 1$  we obtain the following linearized equation:

$$-\frac{d}{dt}\delta\psi(t) + B(t)\delta\psi(t) + C(t-\tau)\delta\psi(t-\tau) + w(t) = 0,$$
(4)

where B and C are the Jacobi matrices of the linearisation [1],  $w(t) = (\xi \operatorname{Re} \eta(t), \xi \operatorname{Im} \eta(t), 0, 0)^T$ . The homogeneous system (4) with  $w(t) \equiv 0$  has two linearly independent periodic solutions, the so-called neutral modes  $\delta \psi_{0\theta} = d\psi_0(t)/dt$  and  $\delta \psi_{0\varphi} = (-\operatorname{Im} A_0, \operatorname{Re} A_0, 0, 0)^T$ , which correspond to the time shift and phase shift symmetry of the system (1)-(3), respectively. Next, we project the increment  $\psi(T_0) - \psi(0) = \delta \psi(T_0) - \delta \psi(0)$  of the perturbed solution  $\psi(t)$  over the period on the neutral eigenfunction  $\delta \psi_{0\theta}$  [7] using the variation of parameters formula for DDEs [6].

Let  $\delta\psi$  be a solution of homogeneous problem (4) with  $w(t) \equiv 0$  and a row vector  $\delta\psi^{\dagger}(t) = (\delta\psi_1^{\dagger}, \delta\psi_2^{\dagger}, \delta\psi_3^{\dagger}, \delta\psi_4^{\dagger})$  be a solution of the adjoint problem

$$\frac{d}{dt}\delta\psi^{\dagger}(t) + \delta\psi^{\dagger}(t)B(t) + \delta\psi^{\dagger}(t+\tau)C(t) = 0.$$
(5)

For solutions  $\phi$  of (4) and  $\phi^{\dagger}$  of (5) we consider the following bilinear form [5,6]

$$[\phi^{\dagger},\phi](t) = \phi^{\dagger}(t)\phi(t) + \int_{-\tau}^{0} \phi^{\dagger}(t+\theta+\tau)C(t+\theta)\phi(t+\theta)d\theta.$$
 (6)

Adjoint periodic eigenfunctions  $\delta \psi^{\dagger}_{0\theta}$  and  $\delta \psi^{\dagger}_{0\varphi}$  (neutral modes of (5)) are  $T_0$ -periodic and satisfy biorthogonality conditions

$$\left[\delta\psi_{0j}^{\dagger},\delta\psi_{0k}\right] = \delta_{jk},\tag{7}$$

where  $j, k = \{\theta, \varphi\}.$ 

Timing jitter  $\sigma_{var}$  is given by the variance of the pulse repetition period fluctuations of the solution  $\delta\psi(t)$  projected on the neutral eigenfunction  $\delta\psi_{0\theta}$  corresponding to the time shift invariance of the model equations. Using variation of constants formula [6] we obtain

$$\sigma_{var}^2 = \operatorname{var}\left(\int_0^{T_0} \delta\psi_{0\theta}^\dagger(s) w(s) ds\right).$$

For the Langevin term w(t), this expression results in the stochastic Ito integral

$$\begin{split} \sigma_{var}^2 &= \xi^2 \mathrm{var} \bigg( \int_0^{T_0} \delta \psi_{0\theta,1}^\dagger(s) dW_1(s) + \\ &\int_0^{T_0} \delta \psi_{0\theta,2}^\dagger(s) dW_2(s) \bigg), \end{split}$$

where  $W_1$  and  $W_2$  are two independent Wiener processes. Since the expected value of this integrals is 0, finally, we obtain

$$\sigma_{var}^2 = \xi^2 \int_0^{T_0} \left( \left( \delta \psi_{0\theta,1}^\dagger(s) \right)^2 + \left( \delta \psi_{0\theta,2}^\dagger(s) \right)^2 \right) ds.$$
(8)

Fig. 1(b) indicates that the timing jitter values estimated with the help of asymptotic formula (8) are in good agreement with those obtained numerically from the field amplitude time traces [12, 13].



Figure 1: (a) Time traces of the mode-locked solution of Eqs. (1)-(3) at  $\xi = 0$  (solid) and  $\xi = 0.05$  (dashed). (b) Comparison of results obtained using the asymptotic formula (8) (line) and those of numerical pulse timing jitter calculation from the amplitude time traces (circles) for different values of noise strength  $\xi$ . Other parameters are  $\tau = 25$  ps,  $\kappa = 0.3$ ,  $\gamma^{-1} = 125$  fs,  $\gamma_g^{-1} = 500$  ps,  $\gamma_q^{-1} = 5$  ps,  $q_0^{-1} = 10$  ps,  $g_0^{-1} = 250$  ps,  $\alpha_q = 1$ ,  $\alpha_g = 2$ , s = 50.

## 3 Experiment

The experiments were performed with two-section monolithic quantum-dot mode-locked devices. The active region consisted of 15 layers of InAs quantum-dots grown on GaAs substrate at Innolume GmbH. The devices were cleaved with no coatings applied to the facets and mounted on a temperature controlled stage. We used a set of lasers with different absorber section lengths in the experiments.

Pulse timing jitter fluctuations were measured via the integration of the normalized power spectral density (PSD),  $L_{RF}(f)$ , over a certain range [21]:

$$\sigma^{i}(f_{1}, f_{2}) = \frac{T_{r}}{2\pi} \sqrt{2 \int_{f_{1}}^{f_{2}} L_{RF}(f) df},$$
(9)

where  $\sigma^i$  is an integrated timing jitter,  $f_1$  and  $f_2$  are the lower and upper integration limits, and  $T_r$  is the pulse train period.

The single sideband PSD was measured from the output radio-frequency (RF) spectrum using an Advantest electronic spectrum analyzer (ESA), a fast photodetector, and an amplifier. The

normalized PSD is given by the expression:

$$L_{RF}(f) = \frac{S_{RF}(f)}{RBW \times S_t},\tag{10}$$

where  $S_t$  is the peak signal power, RBW is the resolution bandwidth of the ESA, and  $S_{RF}(f)$  is PSD.

The single sideband PSD was integrated over the range of 20 kHz-80 MHz as shown in Fig. 2(a), from both sides. When the RF spectra fit well with the Lorentzian (Fig. 2(a), red) the PSD can be estimated from the Lorentzian fit and multiplied by two.

For the Lorentzian line shape, the integrated timing jitter can be determined from the RF linewidth,  $\Delta \nu_{RF,1}$ , of the first harmonic in the RF spectrum and pulse train period  $T_r$  [10]:

$$\sigma^{i}(f_{1}, f_{2}) = \frac{T_{r}\sqrt{\Delta\nu_{RF,1}}}{2\pi^{3/2}}\sqrt{\frac{1}{f_{1}} - \frac{1}{f_{2}}}.$$
(11)

Fig. 2(b) shows measured timing jitter for the lasers  $P_1$  (red) and  $P_2$  (black) with 17% and 12% absorber sections, respectively, as a function of gain current. For the laser  $P_1$  the jitter demonstrated monotonic decrease with the gain current, while for the laser  $P_2$  the integrated jitter behavior was non-monotonic with values ranging from 11 to 23 ps.



Figure 2: (a) Integration of timing jitter. Measured RF signal (black) and Lorentzian fit (red), laser  $P_2$ . (b) Integrated timing jitter for lasers  $P_1$  (red) and  $P_2$  (black) versus gain current.

Fig. 3(a) shows measured RF spectrum evolution versus gain current. The transition between two mode-locked regimes can be seen at the gain current of around 209 mA with a shift of the repetition rate and decrease of jitter. Examples of the RF spectra at the transition point are shown in Fig. 3(b) by black and red lines for the gain currents of 209 mA and 210 mA, respectively. It can be seen that the RF linewidth after the transition point is about twice narrower.



Figure 3: (a) Evolution of the RF spectrum with gain current. (b) RF spectra for the transition point for the gain current of 209 mA (black) and 210 mA (red). Laser  $P_3$ , reverse bias -4.0 V.

## 4 Numerical results

For our numerical analysis we chose the parameters of model equations (1)-(3) to match those of a 10 GHz laser [15] :  $\tau = 100 ps$ ,  $\gamma^{-1} = 0.5 ps$ ,  $\gamma^{-1}_g = 500 ps$ ,  $\gamma^{-1}_q = 10 ps$ ,  $q^{-1}_0 = 5.56 ps$ ,  $\alpha_q = 1$ ,  $\kappa = 0.3$ , and the level of noise  $\xi = 0.05$  [13].

It has been previously noted that high asymmetry in the phase-amplitude coupling in the gain and absorber sections can be strongly linked to complex dynamical behaviour of semiconductor lasers [20]. This strong coupling can be accounted for by assuming that the difference between linewidth enhancement factors  $\alpha_g - \alpha_q$  in the gain and absorber sections is sufficiently large [15]. We chose the pumping parameter  $g_0$  as the continuation parameter to obtain bifurcation diagrams in Fig. 4 with the help of the DDE-BIFTOOL package [4]. The timing jitter was estimated using Eq. (8) by implementing a MATLAB code for calculation of the eigenfunctions of the adjoint eigenproblem (5).

Motivated by the timing jitter comparison performed in the experiment (see Fig. 3) for the bistable branches of fundamental mode-locked regimes with different pulse repetition frequencies, we find that a bistability between two mode-locked regimes exists in the model (1)-(3) for  $q_0$  between 6.058 and 6.06 ns<sup>-1</sup>, see Fig. 4(a). We see from Fig. 4 (b) that for each of the two branches the timing jitter first decreases with increasing  $g_0$  and then increases towards the instability threshold of the mode-locking regime. Similarly to the experimental results of Fig. 3, timing jitter drops abruptly when the solution switches from the branch staring at the point A to the branch starting at the point B. When the saturation parameter s is increased to 9.5 the fold bifurcations disappear (see dash-dotted line in Fig. 4(a)) and a single branch of mode-locking regime remains stable. However, as it is illustrated by the dash-dotted line in Fig. 4(b) the pulse timing jitter dependence on the pump parameter remains similar to that in the bistable case, demonstrating large peaks near the former instability points. At even higher value of the saturation parameter s = 11 the timing jitter dependence still has a smaller peak in the same region. Our simulations performed with the help of the software package DDE-BIFTOOL indicate that these peaks appear when one of the negative Lyapunov exponents of the mode-locked solution  $\psi_0(t)$  comes very close to zero with the change of the parameter  $g_0$ . Such local increase in the pulse timing jitter obtained numerically for  $\alpha_q = 5$  is in agreement with the experimental results

shown in Fig. 2(b) for the laser  $P_2$ . Finally, for s = 24 we observed monotonous decrease of the pulse timing jitter with the increase of  $g_0$ , which is in agreement with the experimental data obtained for the laser  $P_1$ .



Figure 4: Branches of fundamental ML regime (a) and corresponding timing jitter (b) for s = 24 (dash-dash-dotted), s = 11 (dash-dot-dotted), s = 9.5 (dash-dotted), s = 8.9 (solid and dashed for unstable), where A and B are fold bifurcation points. Other parameters are  $\gamma^{-1} = 0.5$  ps,  $\gamma_q^{-1} = 500$  ps,  $\gamma_q^{-1} = 10$  ps,  $q_0^{-1} = 5.56$  ps,  $\alpha_q = 1$ ,  $\alpha_g = 5$ .

## 5 Conclusion

We have studied theoretically and experimentally the effect of noise on the characteristics of a two-section semiconductor laser operating in fundamental mode-locking regime. We have proposed a semianalytical method for estimation of timing jitter in a system of DDEs (1)-(3) describing this laser. The proposed method can be applied to study noise characteristics of other multimode laser devices modeled by systems of delay-differential equations [13, 14, 16]. Using the software package DDE-BIFTOOL we have studied the dependence of the timing jitter on the injection current and other laser parameters and, in contrast to previous theoretical results [22], we have demonstrated that this dependence can be non-monotonous. Specifically, we have observed a peak in the gain dependence curve of timing jitter, which is related to the presence of a real Lyapunov exponent approaching zero from below. Our results suggest that in the bistable regime of operation the branch of mode-locking regime that is stable at lower injection currents exhibits higher level of pulse timing jitter. A sudden drop in the timing jitter is observed when the laser switches to another branch of mode-locking regime with the increase of the injection current.

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